

RADAR CODE OPTIMIZATION FOR MOVING TARGET DETECTION

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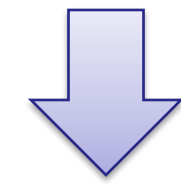
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The discrete-time received signal \mathbf{r} for the range-cell corresponding to the time delay τ can be written as

$$\mathbf{r} = \alpha \mathbf{a} \odot \mathbf{p} + \mathbf{a} \odot \mathbf{c} + \mathbf{w}$$

\swarrow code vector (to be designed) \searrow signal-independent interferences
 \searrow clutter component

$$\text{detection problem: } \begin{cases} H_0 : \mathbf{r} = \mathbf{a} \odot \mathbf{c} + \mathbf{w} \\ H_1 : \mathbf{r} = \alpha \mathbf{a} \odot \mathbf{p} + \mathbf{a} \odot \mathbf{c} + \mathbf{w} \end{cases}$$



$$\text{optimal detector: } |\mathbf{r}^H (\mathbf{M} + \mathbf{A} \mathbf{C} \mathbf{A}^H)^{-1} (\mathbf{a} \odot \mathbf{p})|^2 \underset{H_1}{\overset{H_0}{\gtrless}} \eta$$

In cases where target Doppler shift is unknown, we consider the following *average* metric:

$$\text{tr} \{ \mathbf{A}^H (\mathbf{M} + \mathbf{A} \mathbf{C} \mathbf{A}^H)^{-1} \mathbf{A} \mathbf{W} \}$$

where $\mathbf{W} = \mathbb{E}\{\mathbf{p} \mathbf{p}^H\}$.

code optimization problem:

$$\begin{aligned} \max_{\mathbf{A}} \quad & \text{tr} \{ (\mathbf{A}^{-1} \mathbf{M} \mathbf{A}^{-H} + \mathbf{C})^{-1} \mathbf{W} \} \\ \text{subject to} \quad & \text{tr} \{ \mathbf{A} \mathbf{A}^H \} \leq e \end{aligned}$$

$$\mathbf{W} = \mathbf{V} \mathbf{V}^H \quad \mathbf{V} \in \mathbb{C}^{N \times \delta} \text{ full column-rank}$$

$$\mathbf{R} \triangleq \begin{bmatrix} \theta \mathbf{I} & \mathbf{V}^H \mathbf{A}^H \\ \mathbf{A} \mathbf{V} & \mathbf{M} + \mathbf{A} \mathbf{C} \mathbf{A}^H \end{bmatrix} \quad (18)$$

$$\mathbf{U} \triangleq [\mathbf{I}_\delta \quad \mathbf{0}_{N \times \delta}]^T$$

$$\min_{\mathbf{a}} \quad \mathbf{a}^H ((\mathbf{Y}_2 \mathbf{Y}_2^H) \odot \mathbf{C}^T) \mathbf{a} + 2\Re(\mathbf{d}^H \mathbf{a}) \quad (22)$$

$$\text{subject to} \quad \mathbf{a}^H \mathbf{a} \leq e$$

$$\mathbf{Y} \triangleq [\mathbf{Y}_1 \delta \times \delta \quad \mathbf{Y}_2 N \times \delta]^T \text{ and } \mathbf{d} \triangleq \text{diag}(\mathbf{V}^* \mathbf{Y}_1^* \mathbf{Y}_2^T)$$

CADCODE framework

converges to a stationary point

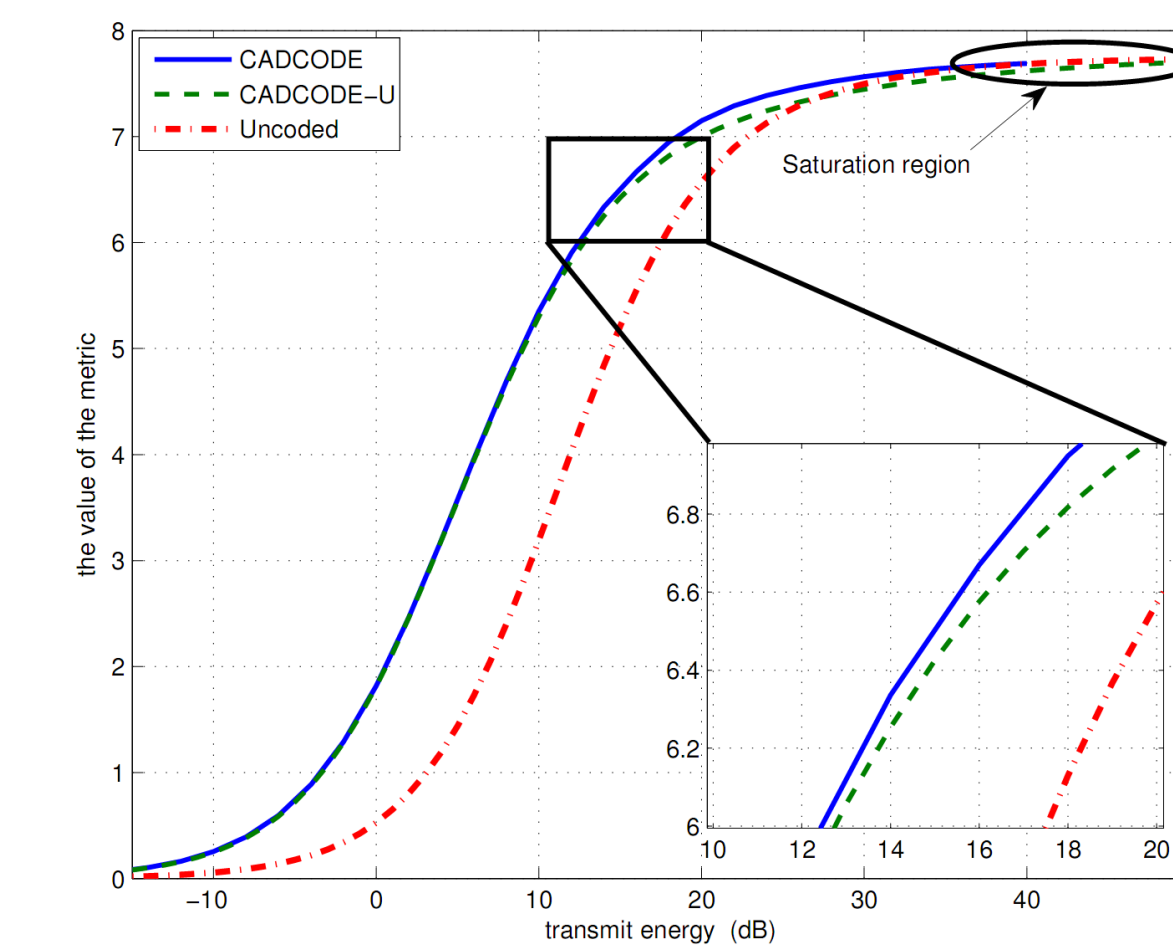
Table 1. CADCODE for Optimal Radar Code Design

Step 0: Initialize the code vector \mathbf{a} using a random vector in \mathbb{C}^N , and form \mathbf{R} as defined in (18).

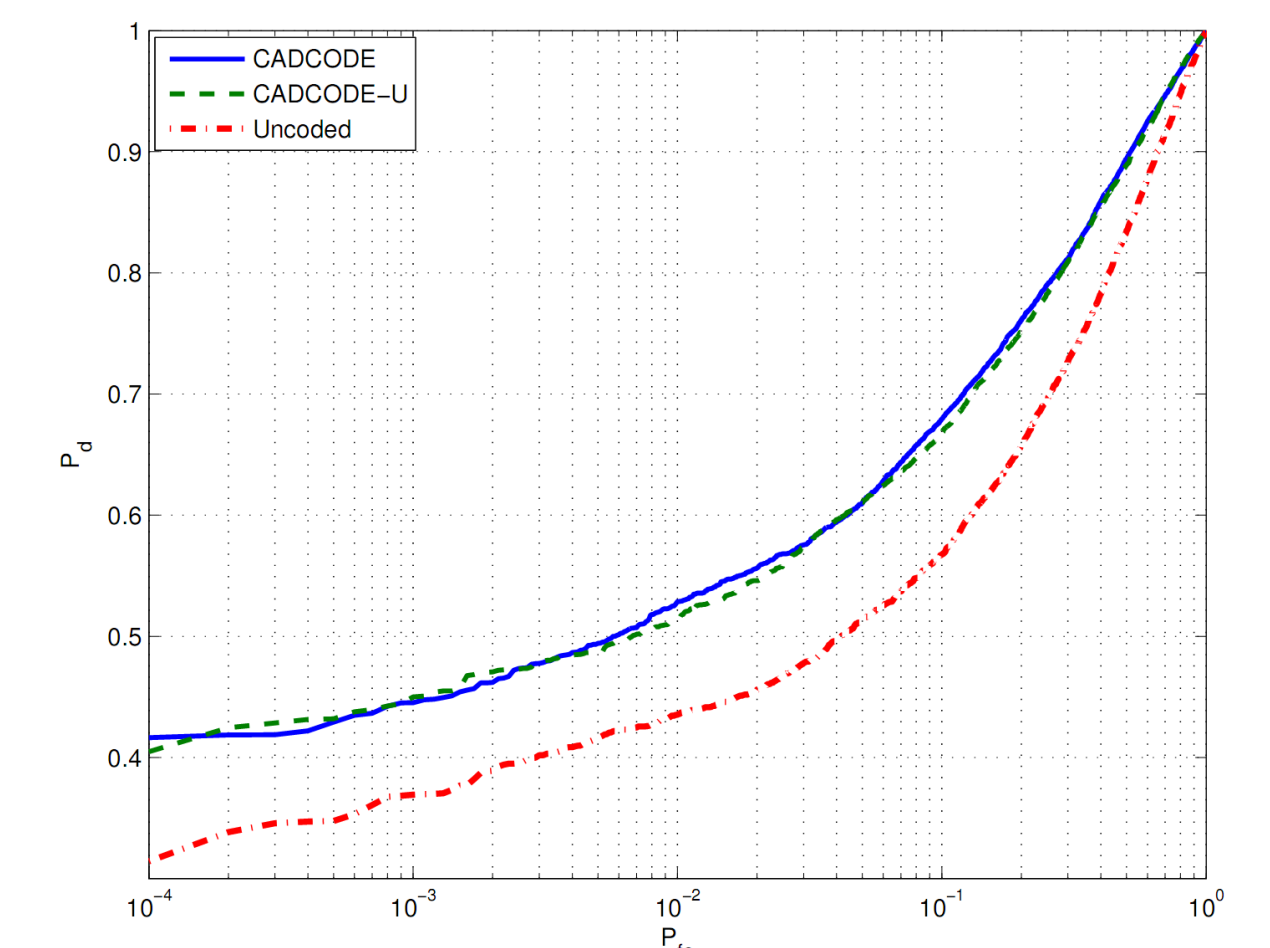
Step 1: Compute $\mathbf{Y} = \mathbf{R}^{-1} \mathbf{U} (\mathbf{U}^H \mathbf{R}^{-1} \mathbf{U})^{-1}$.

Step 2: Solve the optimization problem (22) to obtain the code vector \mathbf{a} .

Step 3: Repeat steps 1 and 2 until a pre-defined stop criterion is satisfied, e.g. $\|\mathbf{a}^{(k+1)} - \mathbf{a}^{(k)}\| \leq \epsilon$ for some $\epsilon > 0$.



(a)



(b)

Fig. 1. The design of optimal codes of length $N = 16$. (a) depicts the values of the metric for CADCODE and CADCODE-U methods as well as the uncoded system vs. the transmit energy. (b) plots the ROC of the optimal detector (unknown ω) for $\sigma^2 = 3$ and transmit energy = 10.

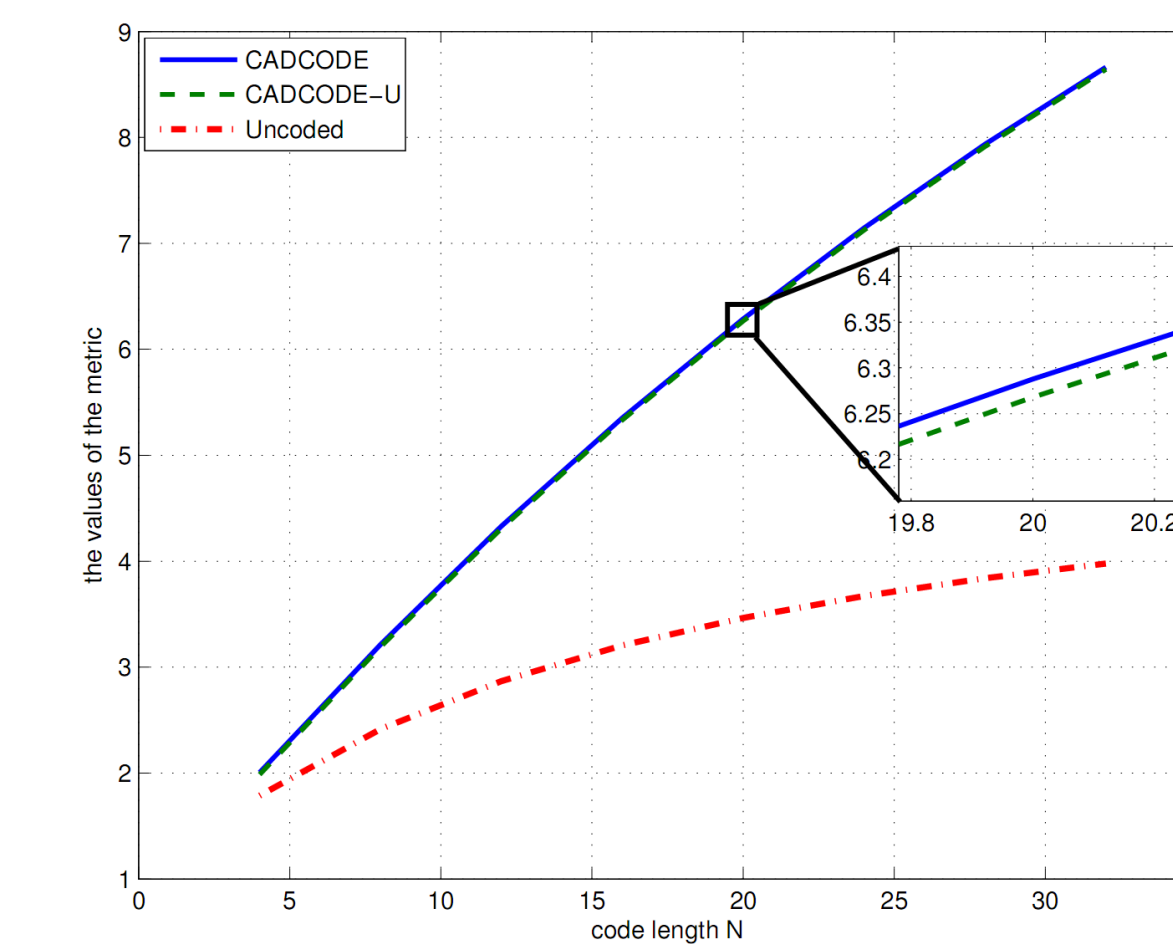


Fig. 2. The values of the metric associated with CADCODE, CADCODE-U, and the uncoded system vs. the code length N .

NUMERICAL EXAMPLES

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