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MAJORIZATION-MINIMIZATION TECHNIQUE FOR MULTI-STATIC RADAR CODE DESIGN

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Problem Formulation: Data Modeling

We consider a multi-static pulsed-radar with one transmitter and N_r widely separated receive antennas. The baseband transmit signal can be formulated as

$$s(t) = \sum_{n=1}^{N} a_n \phi(t - [n-1]T_P)$$
 (1)

where $\phi(t)$ is the basic unit-energy transmit pulse, T_P is the pulse repetition period, N is the number of transmitted pulses, and $\{a_n\}_{n=1}^N$ are the deterministic coefficients that are to be "optimally" determined. The vector $\mathbf{a} \triangleq [a_1 \ a_2 \ \dots \ a_N]^T$ is referred to as the code vector of the radar system.



Problem Formulation: Data Modeling

ullet Therefore, the discrete-time signal corresponding to a certain radar cell for the k^{th} receiver can be described as:

$$\mathbf{r}_k \triangleq \mathbf{s}_k + \mathbf{c}_k + \mathbf{w}_k = \alpha_k \mathbf{a} + \widetilde{\rho}_k \mathbf{a} + \mathbf{w}_k$$
 (6)

binary hypothesis problem

$$\begin{cases}
H_0: & \mathbf{r} = \mathbf{c} + \mathbf{w} \\
H_1: & \mathbf{r} = \mathbf{s} + \mathbf{c} + \mathbf{w}
\end{cases}$$
(7)

$$\mathbf{D}_{k} \triangleq (\sigma_{\mathrm{c},k}^{2} \mathbf{a} \mathbf{a}^{H} + \mathbf{M}_{k})^{-\frac{1}{2}}$$
 $\mathbf{x}_{k} = \mathbf{D}_{k} \mathbf{r}_{k}$
 $\theta_{k} \triangleq \mathbf{a}^{H} \mathbf{D}_{k} \mathbf{x}_{k} / \|\mathbf{a}^{H} \mathbf{D}_{k}\|_{2}$
 $\lambda_{k} = \sigma_{k}^{2} \mathbf{a}^{H} (\sigma_{\mathrm{c},k}^{2} \mathbf{a} \mathbf{a}^{H} + \mathbf{M}_{k})^{-1} \mathbf{a}$

detection problem

$$T = \sum_{k=1}^{N_r} \frac{\lambda_k |\theta_k|^2}{1 + \lambda_k} \underset{\mathcal{H}_1}{\overset{\mathcal{H}_0}{\leqslant}} \eta$$



From KL-Divergence to the Design Problem

$$\mathcal{D}\left(f(\mathbf{r}|\mathbf{H}_0)\|f(\mathbf{r}|\mathbf{H}_1)\right) = \sum_{k=1}^{N_r} \left\{\log(1+\lambda_k) - \lambda_k/(1+\lambda_k)\right\}$$

$$\max_{\mathbf{a},\lambda_k} \quad \sum_{k=1}^{N_r} \left\{\log(1+\lambda_k) - \lambda_k/(1+\lambda_k)\right\}$$
subject to
$$\lambda_k = \sigma_k^2 \mathbf{a}^H (\sigma_{c,k}^2 \mathbf{a} \mathbf{a}^H + \mathbf{M}_k)^{-1} \mathbf{a}$$

$$\max_{n=1,\dots,N} \left\{|a_n|^2\right\} \le \zeta\left(e/N\right)$$

$$\|\mathbf{a}\|_2^2 = e,$$
total transmit energy

PAR-constrained code design

$$PAR(\mathbf{a}) = \max_{n} \{|a_n|^2\}/(\frac{1}{N}\|\mathbf{a}\|_2^2) \le \zeta$$



& The MaMi!

 $\min_{\mathbf{z}} \ \widetilde{f}(\mathbf{z})$ subject to $c(\mathbf{z}) \leq 0$

• Majorization Step: Finding $p^{(l)}(\mathbf{z})$ such that its minimization is simpler than that of $\widetilde{f}(\mathbf{z})$, and that $p^{(l)}(\mathbf{z})$ majorizes $\widetilde{f}(\mathbf{z})$, i.e.

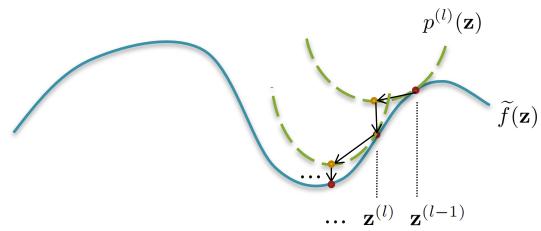
$$p^{(l)}(\mathbf{z}) \ge \widetilde{f}(\mathbf{z}), \quad \forall \mathbf{z} \text{ and } p^{(l)}(\mathbf{z}^{(l-1)}) = \widetilde{f}(\mathbf{z}^{(l-1)})$$
 (13)

with $\mathbf{z}^{(l-1)}$ being the value of \mathbf{z} at the $(l-1)^{th}$ iteration.

• Minimization Step: Solving the optimization problem,

$$\min_{\mathbf{z}} \ p^{(l)}(\mathbf{z}) \quad \text{subject to} \quad c(\mathbf{z}) \le 0 \tag{14}$$

to obtain $\mathbf{z}^{(l)}$.





The Final Form: QP@ Each Iteration.

$$\frac{(l+1)^{th}}{\text{iteration}}$$
:

$$\min_{\mathbf{a}} \quad \mathbf{a}^{H} \left(\sum_{k=1}^{N_{r}} \phi_{k}^{(l)} \mathbf{M}_{k}^{-1} \right) \mathbf{a} - \operatorname{Real} \left(\sum_{k=1}^{N_{r}} \mathbf{a}^{H} \mathbf{d}_{k}^{(l)} \right)$$
subject to
$$\max_{n=1,\dots,N} \left\{ |a_{n}|^{2} \right\} \leq \zeta \left(e/N \right) \tag{25}$$

$$\|\mathbf{a}\|_{2}^{2} = e$$

$$\mathbf{d}_{k}^{(l)} \triangleq (\psi_{k}^{l} / \sqrt{y_{k}^{(l)}}) \mathbf{M}_{k}^{-1} \mathbf{a}^{(l)}$$

$$\phi_{k}^{(l)} \triangleq \frac{\beta_{k}}{1 + \beta_{k} y_{k}^{(l)}} + \beta_{k} (1 + \gamma_{k}) + \frac{\gamma_{k}}{(1 + \lambda_{k}^{(l)})^{2}} \left(\frac{\beta_{k}}{(1 + \beta_{k} y_{k}^{(l)})^{2}} \right)$$

$$\psi_{k}^{(l)} \triangleq \sqrt{y_{k}^{(l)}} \left(\frac{2\beta_{k} (1 + \gamma_{k})}{1 + \beta_{k} y_{k}^{(l)} (1 + \gamma_{k})} + 2\beta_{k} (1 + \gamma_{k}) \right)$$

$$\gamma_{k} = \frac{\sigma_{k}^{2}}{\sigma_{c,k}^{2}}$$

$$\beta_{k} = \sigma_{c,k}^{2}$$



QP with PAR Constraint?

$$\max_{\mathbf{a}} \quad \widehat{\mathbf{a}}^H \mathbf{K} \, \widehat{\mathbf{a}}$$
 subject to
$$\max_{n=1,\cdots,N} \{|a_n|^2\} \le \zeta \, (e/N)$$

$$\|\mathbf{a}\|_2^2 = e$$

$$\widehat{\mathbf{a}} = [\mathbf{a} \ 1]^T, \mathbf{K} = \mu \mathbf{I}_{N+1} - \mathbf{J},$$

$$\mathbf{J} = \begin{bmatrix} \left(\sum_{k=1}^{N_r} \phi_k^{(l)} \mathbf{M}_k^{-1} \right) & -0.5 \left(\sum_{k=1}^{N_r} \mathbf{d}_k^{(l)} \right) \\ -0.5 \left(\sum_{k=1}^{N_r} \mathbf{d}_k^{(l)} \right)^H & 0 \end{bmatrix}$$

 $\mu > \mu_{max} \longrightarrow \text{maximum eigenvalue}$



Power Method-Like Iterations

The code vector \mathbf{a} at the $(l+1)^{th}$ iteration of MaMi can be obtained from $\mathbf{a}^{(p)}$ (at convergence), using the power method-like iterations [11]:

$$\max_{\mathbf{a}^{(p+1)}} \quad \|\mathbf{a}^{(p+1)} - \breve{\mathbf{a}}^{(p)}\|$$
subject to
$$\max_{n=1,\dots,N} \{|a_n^{(p+1)}|^2\} \le \zeta(e/N)$$

$$\|\mathbf{a}^{(p+1)}\|_2^2 = e$$

$$(27)$$

where $\check{\mathbf{a}}^{(p)}$ represents the vector containing the first N entries of \mathbf{K} $\widehat{\mathbf{a}}^{(p)}$.



Algorithm: Summary

Table 1. The MaMi Algorithm for maximizing the KL-divergence with a PAR constraint

Step 0: Initialize **a** with a random vector in \mathbb{C}^N and set the iteration number l to 0.

Step 1: Solve the problem in (25) iteratively considering the nearest-vector problem in (27); set $l \leftarrow l + 1$.

Step 2: Compute $\phi_k^{(l)}$ and $\mathbf{d}_k^{(l)}$.

Step 3: Repeat steps 1 and 2 until a pre-defined stop criterion is satisfied, e.g. $\|\mathbf{a}^{(l+1)} - \mathbf{a}^{(l)}\|_2 \le \xi$ for some $\xi > 0$.



Simulation Results

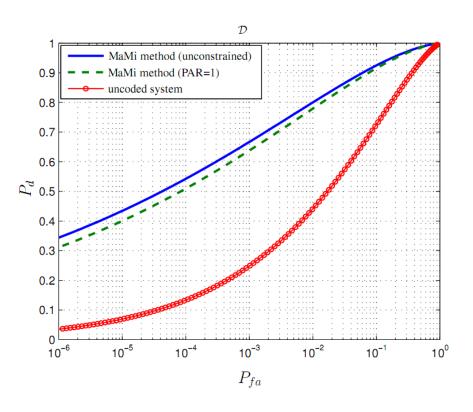


Fig. 1. ROCs of optimally coded and the uncoded systems.



Simulation Results

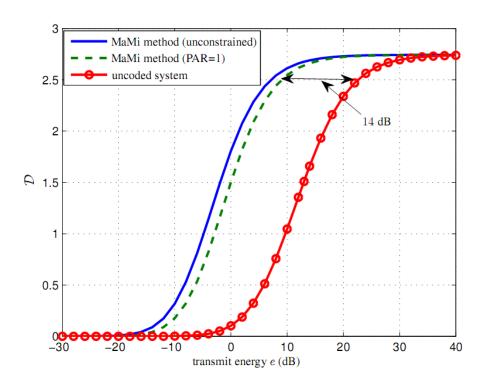


Fig. 2. Behavior of KL-divergence versus transmit energy e for the coded and uncoded systems.



Thank you for your kind attention! -Qs?